## Digital Biquad Design by Magnitude Requirements

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We consider the problem of finding the coefficients  $b_0, b_1, b_2, a_0, a_1, a_2$  of a discrete time biquad transferfunction of the general form:

$$H(z) = \frac{b_0 + b_1 z^{-1} + b_2 z^{-2}}{a_0 + a_1 z^{-1} + a_2 z^{-2}} \tag{1}$$

given 5 requirements on the magnitude response. These requirements are given as pairs of a normalized radian frequency  $\omega_n$  and a desired magnitude-response value  $g_n$  which the filter should have at  $\omega_n$  such that  $g_n = |H(e^{j\omega_n})|$ , where  $n = 1, \ldots, 5$ . As we shall see later, one of the biquad coefficients is actually redundant and may be set to unity. For implementation purposes, it makes most sense to normalize the transfer-function such that  $a_0 = 1$ . We will take care of this (re)normalization later - for the design algorithm, we have to take  $a_0$  as variable.

## From Requirements to a System of Equations

The magnitude-squared response of a general biquad transfer-function (1) is given by:

$$g^{2}(\omega) \hat{=} |H(e^{j\omega})|^{2} = \frac{b_{0}^{2} + b_{1}^{2} + b_{2}^{2} + 2b_{1}(b_{0} + b_{2})\cos(\omega) + 2b_{0}b_{2}\cos(2\omega)}{a_{0}^{2} + a_{1}^{2} + a_{2}^{2} + 2a_{1}(a_{0} + a_{2})\cos(\omega) + 2a_{0}a_{2}\cos(2\omega)}$$
(2)

We define:

$$B_0 = \hat{b}_0^2 + \hat{b}_1^2 + \hat{b}_2^2, \quad B_1 = \hat{b}_1(\hat{b}_0 + \hat{b}_2), \quad B_2 = \hat{b}_0 \hat{b}_2$$
  

$$A_0 = \hat{a}_0^2 + \hat{a}_1^2 + \hat{a}_2^2, \quad A_1 = \hat{a}_1(\hat{a}_0 + \hat{a}_2), \quad A_2 = \hat{a}_0 \hat{a}_2$$
(3)

We have 5 magnitude-requirements at 5 normalized radian frequencies given as pairs  $(\omega_n, g_n)$ , n = 1, ..., 5. Because we will mostly deal with the squared magnitude, we'll assign a name to that squared magnitude by defining  $p_n = g_n^2$  such that may now deal with the 5 pairs  $(\omega_n, p_n)$ . Using the definitions (3) in (2), these 5 requirements lead to the system of 5 equations:

$$p_n = \frac{B_0 + 2B_1 \cos(\omega_n) + 2B_2 \cos(2\omega_n)}{A_0 + 2A_1 \cos(\omega_n) + 2A_2 \cos(2\omega_n)} \qquad n = 1, \dots, 5$$
(4)

Defining furthermore:

$$u_n = 2\cos(\omega_n), \qquad v_n = 2\cos(2\omega_n) \tag{5}$$

we may simplify this system to:

$$p_n = \frac{B_0 + B_1 u_n + B_2 v_n}{A_0 + A_1 u_n + A_2 v_n} \qquad n = 1, \dots, 5$$
(6)

We now fix  $A_0 = 1$  and multiply both sides by the denominator - we get:

$$p_n(1 + A_1u_n + A_2v_n) = p_n + p_nA_1u_n + p_nA_2v_n = B_0 + B_1u_n + B_2v_n \qquad n = 1, \dots, 5$$
(7)

Bringing the terms containing  $A_1, A_2$  over to the right hand side, this becomes:

$$p_n = B_0 + B_1 u_n + B_2 v_n - p_n A_1 u_n - p_n A_2 v_n \qquad n = 1, \dots, 5$$
(8)

This is a linear system of 5 equations for 5 unknowns  $B_0, B_1, B_2, A_1, A_2$ . We can also write this in matrix form:

$$\begin{pmatrix} 1 & u_1 & v_1 & -p_1u_1 & -p_1v_1 \\ 1 & u_2 & v_2 & -p_2u_2 & -p_2v_2 \\ 1 & u_3 & v_3 & -p_3u_3 & -p_3v_3 \\ 1 & u_4 & v_4 & -p_4u_4 & -p_4v_4 \\ 1 & u_5 & v_5 & -p_5u_5 & -p_5v_5 \end{pmatrix} \begin{pmatrix} B_0 \\ B_1 \\ B_2 \\ A_1 \\ A_2 \end{pmatrix} = \begin{pmatrix} p_1 \\ p_2 \\ p_3 \\ p_4 \\ p_5 \end{pmatrix}$$
(9)

This system may now be solved for our intermediate variables  $B_0, B_1, B_2, A_1, A_2$  by standard techniques for linear systems (such as Gaussian elimination).

## **Computing the Biquad Coefficients**

Having obtained our intermediate variables  $B_0, B_1, B_2, A_1, A_2$ , the next task is to go back to the actual biquad coefficients  $b_0, b_1, b_2, a_0, a_1, a_2$ . From the first line of (3), we see:

$$B_2 = b_0 b_2 \Leftrightarrow b_2 = \frac{B_2}{b_0}, \qquad B_1 = b_1 (b_0 + b_2) \Leftrightarrow b_1 = \frac{B_1}{b_0 + b_2} = \frac{B_1}{b_0 + B_2/b_0}$$
 (10)

and so:

$$B_0 = b_0^2 + b_1^2 + b_2^2 = b_0^2 + \left(\frac{B_1}{b_0 + B_2/b_0}\right)^2 + \left(\frac{B_2}{b_0}\right)^2 \tag{11}$$

Bringing  $B_0$  over to the right hand side and evaluating the squares gives:

$$0 = b_0^2 + \frac{B_1^2}{b_0^2 + 2B_2 + B_2^2/b_0^2} + \frac{B_2^2}{b_0^2} - B_0$$
(12)

which we may consider as a root-finding problem for  $b_0^2$ . We must now solve this root-finding problem and take the square-root to obtain  $b_0$  from  $b_0^2$ . Once  $b_0$  is known,  $b_1$  and  $b_2$  may be computed straightforwardly by plugging  $b_0$  back into (10). Empirically, it turns out that the root-finding problem may be solved by Newton iteration using  $b_0^2 = 1$  as start value. The very same procedure can be used to compute  $a_0, a_1, a_2$ from  $A_0 = 1, A_1, A_2$ . [TODO: investigate convergence properties of the Newton iteration more thoroughly]

## A Normalized, Stable, Minimum-Phase Biquad

Our biquad coefficients so obtained do not necessarily lead to a stable, minimum-phase filter nor do they satisfy the  $a_0 = 1$  normalization, which is usually desired for implementation purposes. Transforming the coefficient-set to a stable, minimum-phase set is done by reflecting poles and zeros outside the unit circle into the unit circle. A re-normalization to  $a_0 = 1$  is straightforward by dividing all coefficients by  $a_0$ .